

Lec 19 - circuit switched - isochronous data
 - packet " - bursty "

- HDLC frames start & end w/

HDLC flags: 01111110

bit stuffing: add a '0' after 5 '1's

- OSI 7-layer model:

LLC

MAC

Physical

} DDL

waveform

PHY

error detection & correction

LLC

media access control

MAC

addressing

LLC

SDU - payload

PDU - frame w/ headers & trailers

Lec. 9 FD - tx & rx at same time

HD - one direction at a time

TDD -

TX	RX	TX	RX
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FDD - $f_1 \rightarrow$
 $f_2 \leftarrow$

ASK

BPSK

QPSK

M-QAM

(M=4, 16, 64, 256, ...)

bits/symbol

1

1

2

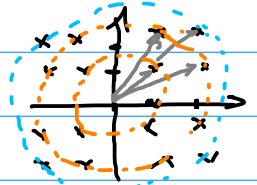
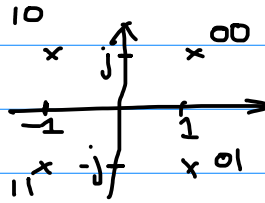
$\log_2 M = 2, 4, 6, 8$
bits/symbol = $\log_2 M$

equation

$$m(t) \cos(\omega_c t)$$

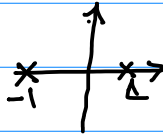
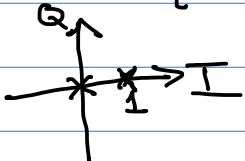
$$m(t) = \begin{cases} 0 \\ 1 \end{cases}$$

=

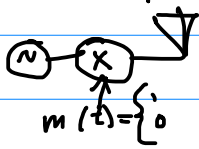


(e.g. 16-QAM)

constellation



mod



$$m(t) = \begin{cases} +1 \\ -1 \end{cases}$$

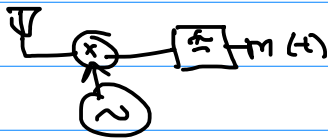
$$m(t) = \pm 1 \pm j$$

$$m(t) = \sqrt{2} e^{j\theta}$$

$$m(t) = \pm 1, \pm 3$$

$$\pm j, \pm 3j$$

demod

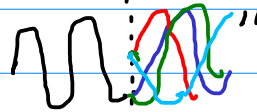
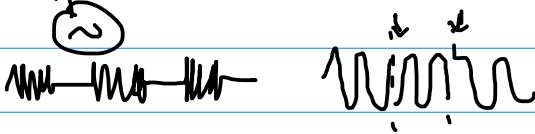


$$\theta = \pm 45^\circ, \pm 135^\circ$$

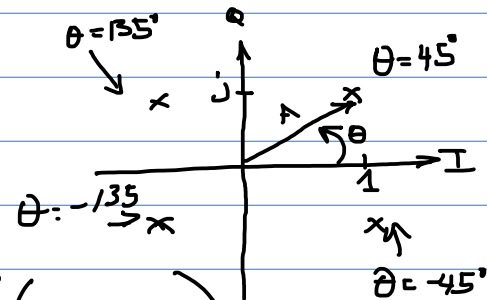
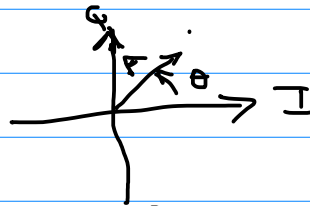
$$s(t) = \cos(\omega_c t)$$

see next question

waveform



$$A e^{j\theta} = A(\cos\theta + j\sin\theta)$$



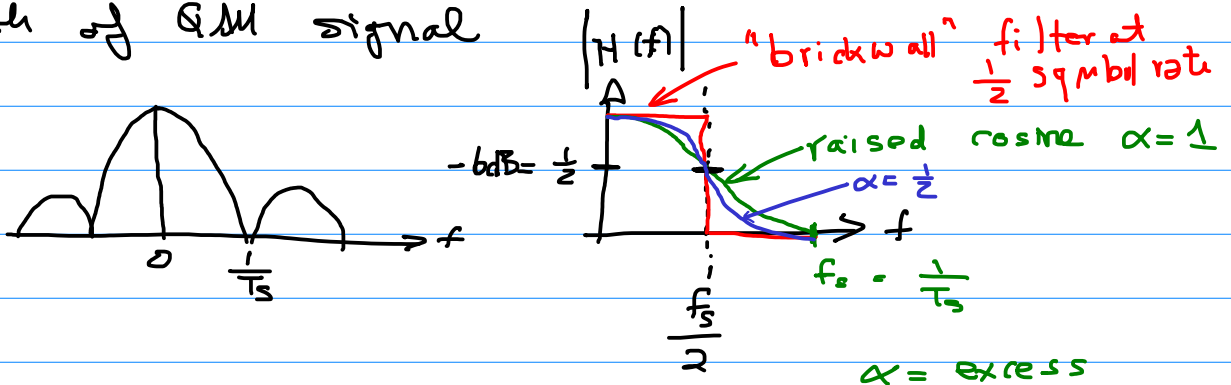
$$\frac{\pi}{4} + \left\{ \frac{\pi}{2}, \frac{3\pi}{2}, \frac{5\pi}{2}, \frac{7\pi}{2} \right\} \sqrt{2} \left(e^{j(\dots)} \right)$$

$$(3) s(t) = \sqrt{2} \cos(\omega_c t + \theta) \quad \theta = \pm 45^\circ, \pm 135^\circ$$

gray codes:

—	0	0	↕	↕
—	0	1	↕	
—	1	0		
—	1	0		

bandwidth of QM signal

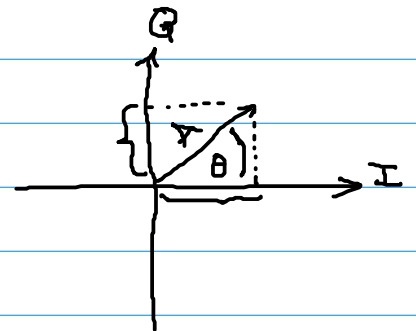


1 Mb/s

$\frac{1}{2}$ voltage = $\frac{1}{4}$ power
= -6 dB

$\frac{1}{\sqrt{2}}$ voltage = $\frac{1}{2}$ power
= -3 dB

for quadrature downconverter
if input is $A \cos(\omega_c t + \theta)$
 $I = A \cos(\theta)$
 $Q = A \sin(\theta)$



	deviation/bitrate	filtering
GMSK	$\frac{1}{2}$	gaussian (t or f)
MSK	$\frac{1}{2}$	any
FSK	any	any

Lec. 11

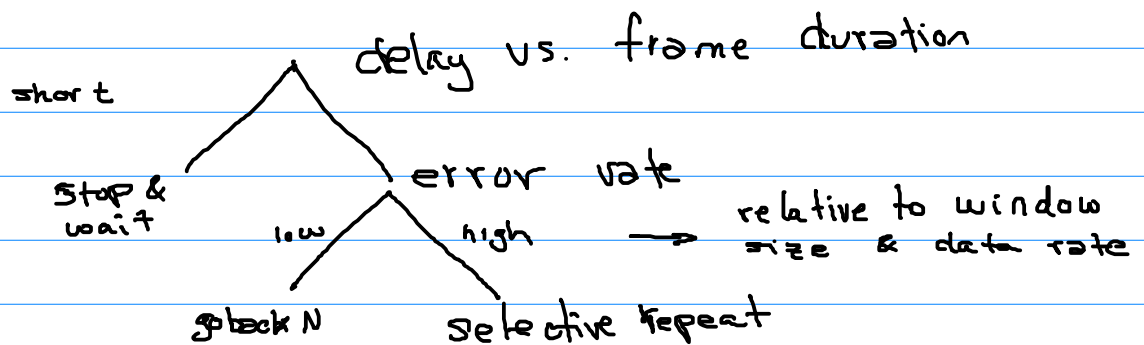
$$C = B \log_2 \left(1 + \frac{SNR}{N} \right) \text{ Shannon Capacity}$$

bandwidth \swarrow SNR in linear units \swarrow

if operating at $> C \Rightarrow$ error rate cannot be reduced to a low value

$< C \Rightarrow$ BER ^{it more} can be reduced to arbitrarily low value

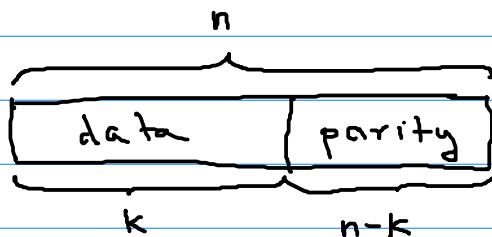
ARQ Selection



even parity: even number of 1's

0110 1011 \rightarrow 5 '1' bits (odd number)

for even parity, parity = 1
odd " , " = 0



(n,k) code

e.g. $(7,4)$

$n=7$ total bits

$k=4$ data bits

Hamming Distance = # of bits different between two codewords

eg

1011	1100	
1101	1101	
✓✓	✓	← 3 bits different

(3,1)	possible transmitted codewords	if 111 received									
	<table style="margin-left: 20px;"> <tr><td style="padding: 5px;">010</td></tr> <tr><td style="padding: 5px;">101</td></tr> </table>	010	101	<table style="margin-left: 20px;"> <tr><td style="padding: 5px;">2</td></tr> <tr><td style="padding: 5px;">1</td></tr> </table>	2	1					
010											
101											
2											
1											
	<table style="margin-left: 20px;"> <tr> <td style="padding: 5px;"> <table style="border-collapse: collapse;"> <tr><td style="padding: 2px 5px;">111</td></tr> <tr><td style="padding: 2px 5px;">010</td></tr> <tr><td style="border-top: 1px solid black; padding: 2px 5px;">✓✓</td></tr> </table> </td> <td style="padding: 5px 20px;"></td> <td style="padding: 5px;"> <table style="border-collapse: collapse;"> <tr><td style="padding: 2px 5px;">111</td></tr> <tr><td style="padding: 2px 5px;">101</td></tr> <tr><td style="border-top: 1px solid black; padding: 2px 5px;">✓</td></tr> </table> </td> </tr> </table>	<table style="border-collapse: collapse;"> <tr><td style="padding: 2px 5px;">111</td></tr> <tr><td style="padding: 2px 5px;">010</td></tr> <tr><td style="border-top: 1px solid black; padding: 2px 5px;">✓✓</td></tr> </table>	111	010	✓✓		<table style="border-collapse: collapse;"> <tr><td style="padding: 2px 5px;">111</td></tr> <tr><td style="padding: 2px 5px;">101</td></tr> <tr><td style="border-top: 1px solid black; padding: 2px 5px;">✓</td></tr> </table>	111	101	✓	
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111											
010											
✓✓											
111											
101											
✓											

$k = \log_2 (\# \text{ of codewords}) = \log_2 (2) = 1$
 $n = \# \text{ of bits in codewords}$

Lec. 12

in GF(2) 0, 1

(+)	add \equiv XOR	<table style="border-collapse: collapse;"> <tr><td style="padding: 0 5px;">+</td><td style="border-left: 1px solid black; padding-left: 5px;">01</td></tr> <tr><td style="padding: 0 5px;">-</td><td style="border-left: 1px solid black; padding-left: 5px;">01</td></tr> <tr><td style="padding: 0 5px;">x</td><td style="border-left: 1px solid black; padding-left: 5px;">1</td></tr> </table>	+	01	-	01	x	1
+	01							
-	01							
x	1							
(x)	multiply \equiv AND							

$1+1=0$

$010 \Rightarrow 0x^2 + 1x^1 + 0x^0$

6 data bits = k
 3 parity bits = n-k
 9 total bits = n

$$M(x) = 010101$$

$$G(x) = 1101$$

$$1x^3 + 1x^2 + 0x + 1$$

$$n-k = 3$$

$$\begin{array}{r}
 \overline{010101} \\
 1101 \overline{) 010101000} \\
 \underline{0000} \\
 1010 \\
 \underline{1101} \\
 1111 \\
 \underline{1101} \\
 0100 \\
 \underline{0000} \\
 1000 \\
 \underline{1101} \\
 1010 \\
 \underline{1101} \\
 111
 \end{array}$$

transmit: 010101111

$$\begin{array}{r}
 \overline{011011} \\
 1101 \overline{) 010101111} \\
 \underline{0000} \\
 1010 \\
 \underline{1101} \\
 1111 \\
 \underline{1101} \\
 0101 \\
 \underline{0000} \\
 1011 \\
 \underline{1101} \\
 1101 \\
 \underline{1101} \\
 000
 \end{array}$$

probability of undetected error: (CRC fails to detect an error)
 for completely random bits:

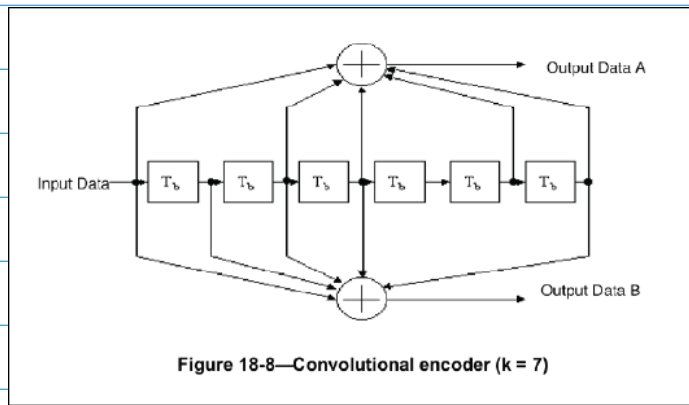
$$\frac{1}{2^{n-k}}$$

$n-k = \# \text{ bits in CRC}$
 (16 or more commonly, 32)

eg for 32-bit CRC = $\frac{1}{2^{32}} \approx 3 \times 10^{-10}$

for ¹ burst error of length $< n-k$
 prob. of undetected error = 0.

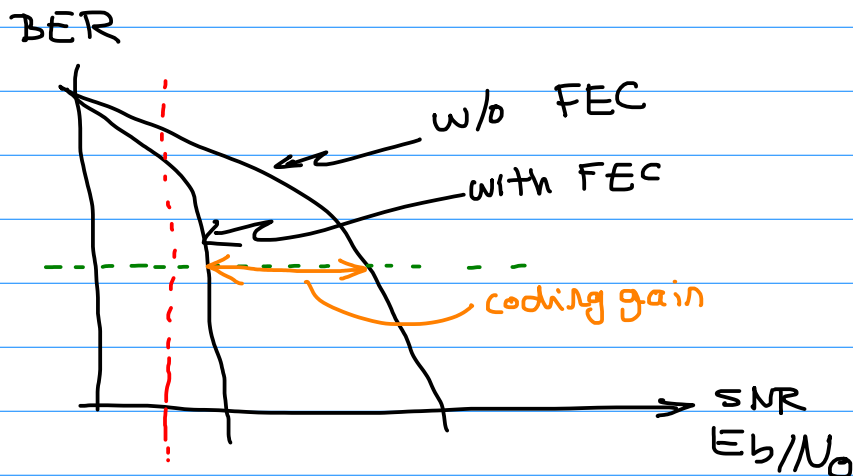
Lec. 13



$$K = \# \text{ bits in } SR + 1 \text{ (input)}$$

$$= 6 + 1 = 7$$

$$R = \text{rate} = \frac{\# \text{ input bits}}{\# \text{ output bits}} = \frac{1}{2} \quad \left(\begin{array}{l} k=1 \\ n=2 \end{array} \right)$$



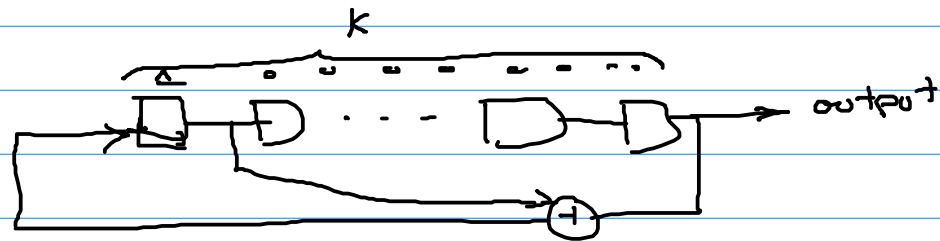
Turbo
LDPC } FEC Codes \Rightarrow allow systems to almost reach Shannon capacity

RS (Reed Solomon) \Rightarrow corrects words instead of bits, \therefore good for bursty errors

Lec. 14

signal	voltage or current varies w/ time
noise	random "
pseudo-random PN	predictable w/ known statistics noise-like "
PRBS	2-valued "
ML PRBS	period $2^k - 1$ " (k is bits of state in generator)

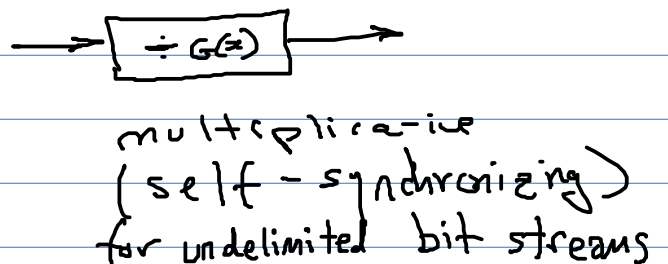
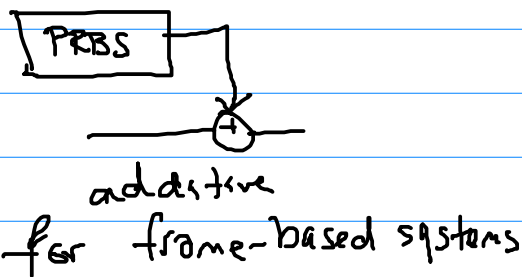
101 101 101 101 $k=2$



period = $(2^k - 1) T_b$

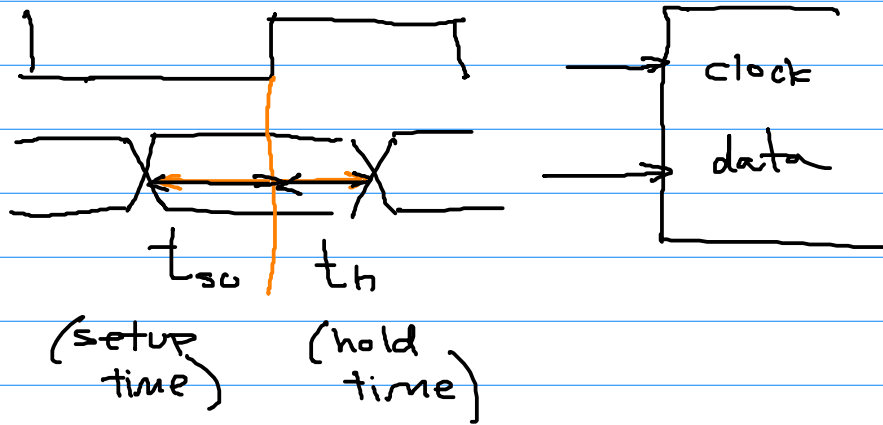
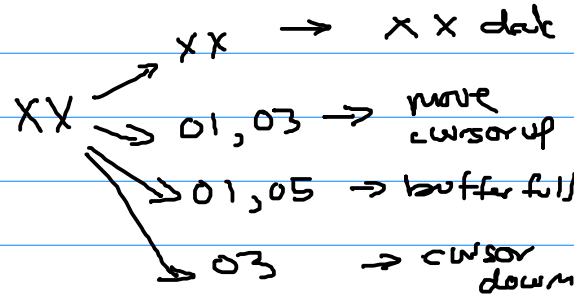
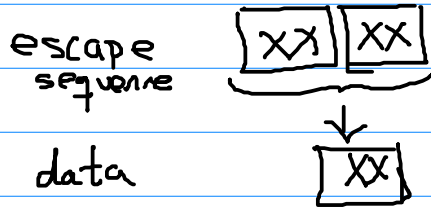
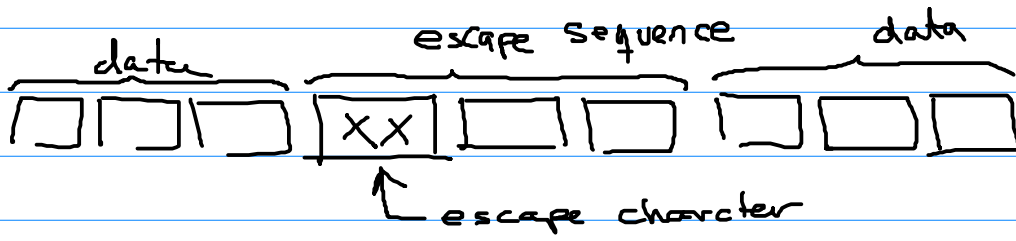
maximum length of a run of 0s is $k-1$

scrambler.



Lec. 15

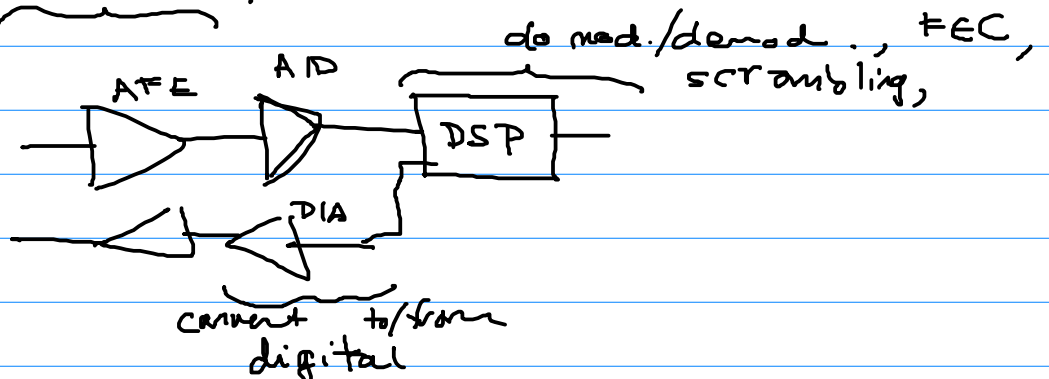
escape sequences



Lec. 17

reason for integration : cost
 " " digital : cost +
 (Moore's law)

filtering
 gain
 conversion to/from baseband



	hardware	software
sampling rate	high	low
algorithmic complexity (how complicated are the algorithms) (lines of code)	low	high
computational complexity # instructions / sample	high	low

full-custom ICs
 semi-custom " (gate arrays)
 FPGA (CPLD's)
 discrete logic ICs

volume
 very high (million)
 ↓ in between
 small (1 → thousand)
 very simple

Lec. 18

- self configuring
- supplies power
- faster
- hot plug
- standard drivers (device classes)