

## System Verilog

*This lecture describes Verilog types, operators and the evaluation of expressions as well as the most common Verilog statements. More details are available in the System Verilog standard, IEEE Std 1800-2012.*

*After this lecture you should be able to: predict the size and value of a Verilog expression that uses the signals, constants and operators described below; and predict the flow of control between the statements described below.*

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### Types and Logic Values

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Verilog 4-state types, such as `logic`, can have four values: 0 (false), 1 (true), `x` (unknown) and `z` (high impedance). These are used for modeling logic.

2-state types such as `bit` and `integer` can have values 0 and 1. These are primarily used as counters and array indices.

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### Numeric Constants

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In addition to declaring the size and base as described previously, constants (“literals”) can also be declared as signed by prefixing the base with ‘s’ (e.g. `4'shf` for a 4-bit -1). Decimal constants are treated as signed by default.

A numeric constant can also include `x` (unknown) or `z` (high-impedance) values. These have useful interpretations for synthesis (don’t-care and high-impedance respectively) but when used in simulations the result, in most cases, will be unknown (`x`).

The notations `'0` and `'1` are convenient abbreviations for a constant that is all-zeros or all-ones.

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### Signedness and Size

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Variables and expressions can have a size (measured in bits) and can be signed or unsigned. The behaviour of some operators depends on the signedness of the operands.

Signedness does not affect the values of the bits (0 or 1) only how operators act on them. Negative values of signed values are assumed to be in two’s-complement form.

Decimal literals are signed and based literals are unsigned unless `s` is used with the base.

The result of an expression is unsigned unless all the operands are signed.

Figure 1<sup>1</sup> shows how the size of an expression depends on its operands.

Values are left-padded or left-truncated as necessary. Padding replicates the sign bit only for signed values.

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### Arrays

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Variables may have multiple “packed” and “unpacked” dimensions.

Packed dimensions are those given before the signal name. These bits are stored contiguously (“packed”) and the packed item can be treated as a scalar – a single number – in expressions. Packed dimensions typically model a word or bit fields within a word (e.g. a 32-bit word composed of 4 bytes of 8 bits).

Unpacked dimensions appear after the signal name. These bits may not necessarily be stored contiguously. Unpacked dimensions model memories where only one element can be accessed at a time.

In array references, the unpacked dimension(s) are specified first, followed by the packed dimensions (if any).

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### Array Literals

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Array literals (constants) can be defined by grouping the individual elements within `{...}`. The quote distinguishes array literal syntax from the syntactically similar concatenation operator.

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### Examples

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The examples below illustrate the rules described above.

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<sup>1</sup>Tables are from the IEEE System Verilog Standard, IEEE Std 1800-2012.

**Table 11-21—Bit lengths resulting from self-determined expressions**

Expression	Bit length	Comments
Unsigned constant number	Same as integer	
Sized constant number	As given	
i op j, where op is: + - * / % &   ^ ^~ ~^	max(L(i),L(j))	
op i, where op is: + - ~	L(i)	
i op j, where op is: === !=== == != > >= < <=	1 bit	Operands are sized to max(L(i),L(j))
i op j, where op is: &&    -> <->	1 bit	All operands are self-determined
op i, where op is: & ~&   ~  ^ ~^ ^~ !	1 bit	All operands are self-determined
i op j, where op is: >> << ** >>> <<<	L(i)	j is self-determined
i ? j : k	max(L(j),L(k))	i is self-determined
{i, ..., j}	L(i)+..+L(j)	All operands are self-determined
{i{j, ..., k}}	i × (L(j)+..+L(k))	All operands are self-determined

Figure 1: Size of Expressions.

```

module ex12 ;
  initial begin

    logic [3:0] x ;
    logic signed [15:0] y ;
    logic [3:0] [7:0] z [15:0] ;

    x = 4'b01xz ; //
    x = -1 + 0 ; // x=15

    y = -1 + 4'shf ; // y=-2
    x = y ; // x=14

    z[0] = '1 ; // z[0]=4294967295
    z[0] = {4{4'b1}} ; // z[0]=4369
    z[0][0][7] = 1 ; // z[0]=4497
    z[15:0] = '{16{z[0]}} ; // z[*]=4497

  end
endmodule

```

**Exercise 1:** What are the packed and unpacked dimensions of each declaration?

**Exercise 2:** What are the signedness, size and value of each constant and each expression above?

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## Operators

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Figure 2 below lists the Verilog operators and Figure 3 their precedence. Operators that differ from those

in C and that are widely supported for synthesis are described below.

**Arithmetic vs Logical Shift** Left shift always zero-fills on the right. Arithmetic right shifts (>>) replicate the sign bit if the result is signed. Logical right shifts (>>) always zero-fill.

**Logical Reduction Operators** These unary (one operand) operators apply a logical operation to the bits of the operand. For example, to test if any bit is set we can apply the or-reduction operator.

**Conditional Operator** The result when the conditional expression contains x or z is not what you might expect (the rules are complex).

**Equality** The regular equality comparison operators (e.g. ==) return x if either operand contains x or z.

The wildcard equality comparison (==?) excludes bits that are x or z in the right operand

**Table 11-1—Operators and data types**

Operator token	Name	Operand data types
=	Binary assignment operator	Any
+= -= /= *=	Binary arithmetic assignment operators	Integral, <b>real</b> , <b>shortreal</b>
%=	Binary arithmetic modulus assignment operator	Integral
&=  = ^=	Binary bitwise assignment operators	Integral
>>= <<=	Binary logical shift assignment operators	Integral
>>>= <<<=	Binary arithmetic shift assignment operators	Integral
?:	Conditional operator	Any
+ -	Unary arithmetic operators	Integral, <b>real</b> , <b>shortreal</b>
!	Unary logical negation operator	Integral, <b>real</b> , <b>shortreal</b>
~ & ~&   ~  ^ ~^ ^~	Unary logical reduction operators	Integral
+ - * / **	Binary arithmetic operators	Integral, <b>real</b> , <b>shortreal</b>
%	Binary arithmetic modulus operator	Integral
&   ^ ^~ ~^	Binary bitwise operators	Integral
>> <<	Binary logical shift operators	Integral
>>> <<<	Binary arithmetic shift operators	Integral
&&    -> <->	Binary logical operators	Integral, <b>real</b> , <b>shortreal</b>
< <= > >=	Binary relational operators	Integral, <b>real</b> , <b>shortreal</b>
=== !=	Binary case equality operators	Any except <b>real</b> and <b>shortreal</b>
== !=	Binary logical equality operators	Any
==? !=?	Binary wildcard equality operators	Integral
++ --	Unary increment, decrement operators	Integral, <b>real</b> , <b>shortreal</b>
<b>inside</b>	Binary set membership operator	Singular for the left operand
<b>dist</b>	Binary distribution operator	Integral
{ } {{}}	Concatenation, replication operators	Integral
{<<{}} {>>{}}	Stream operators	Integral

Figure 2: Verilog operators.

from the comparison. The result is always 0 or 1.

Concatenations of variables can be used on the left hand side of an assignment.

**Concatenation** Bits can be concatenated by separating expressions with commas and surrounding them with braces ({}).

**Replication** The syntax is similar to concatenation but uses two pairs of nested braces and repetition value.

Table 11-2—Operator precedence and associativity

Operator	Associativity	Precedence
() [] :: .	Left	<p>Highest</p> <p>Lowest</p>
+ - ! ~ & ~&   ~  ^ ~^ ^~ ++ -- (unary)		
**	Left	
* / %	Left	
+ - (binary)	Left	
<< >> <<< >>>	Left	
< <= > >= inside dist	Left	
== != === !== ==? !=?	Left	
& (binary)	Left	
^ ~^ ^~ (binary)	Left	
(binary)	Left	
&&	Left	
	Left	
?: (conditional operator)	Right	
-> <->	Right	
= += -= *= /= %= &= ^=  = <<= >>= <<<= >>>= := :/ <=	None	
{ } { }	Concatenation	

Figure 3: Operator Precedence.

**Array Slices** The array subscript operator can be used to extract contiguous portions (slices) of an array. The bit order cannot be reversed.

```

y = x[6:3] === 7'b100x ; // y=1
y = x ==? 8'h4x ; // y=1
y = {x[7:4],x[6]} ; // y=0x0009
end
endmodule

```

**Cast** Although not an operator, a cast ('') can be used to change the type, size or signedness of an expression.

### Examples

```

module ex13 ;
  initial begin

    logic [15:0] x ;
    logic signed [15:0] y ;

    x = 16'hfff0 ; // x=65520
    y = x >>> 1 ; // y=0x7ff8
    y = signed'(x) >>> 1 ; // y=0xffff8
    y = |y ; // y=1

    x = 8'h4x ; // x=X
    y = x == 8'h4x ; // y=X

```

### Modules

Modules represent the top-level blocks of a hardware description. A module includes declarations and parallel (concurrently executing) statements between `module` and `endmodule`. The module's header defines the module's name, parameters and ports. Ports can be `in`, `out` or `inout` (to model bidirectional signals).

When a module is instantiated, a signal can be connected to module port by port order (`sig`); by port name together with an explicit signal name (`.port(sig)`), by port name with the signal name im-

plicit (.port) and using a wildcard (.\*) that matches all remaining matching port and signal names. Expressions can be used instead of a signal name.

## Parameters

Parameters are named constants. They are often included in the module's declaration to allow customization of each instantiation of a module. Parameters such as bus width or clock frequency are common.

Default values can be specified for parameters. As with signals, the values of parameters can be specified by position or by name.

The following example shows how parameters with default values can be specified, how they can be used to specify the dimensions of an array port and how system functions such as \$size(), \$right() and \$clog2(x) ( $\lceil \log_2(x) \rceil$ ) can be used to compute parameter values in module instantiations:

```
module setbits #(M=7,L=0) (output [M:L] x);
    assign x = '1 ;
endmodule

module ex18 ;
    logic [31:16] x ;
    setbits #(.L($right(x,1)),.M(31)) s0(x);
endmodule
```

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## Parallel Statements

A module can contain any number of the following parallel statements, all of which execute concurrently.

### always Procedural Blocks

always blocks execute the following statement in an infinite loop. Execution of the next statement is often controlled by one of the following:

**#number** delays *number* before each execution. This is not synthesizable but is useful for simulation.

**@(expression)** waits until the value of the expression (the “sensitivity list”) changes. This can be used to model combinational, latched or flip-flop logic.

The type of logic generated by the always block depends on the the sensitivity list and which variables are assigned to within the block.

If for some conditions variables are not assigned to within the block then the language semantics require that memory be generated so that the previous value is retained. This memory can be edge-triggered (when the sensitivity list uses **posedge** or **negedge**) or a latch (otherwise). On the other hand, if all variables in the sensitivity list are updated each time the block executes then combinational logic is generated.

A common mistake is to omit signals from the sensitivity list or not assign to a variable. This results in unintended latched logic.

To avoid this, System Verilog has three variants of the always procedural block: **always\_ff**, **always\_comb** and **always\_latch** that document the designer's intent. A warning or error is generated if the sensitivity list or assignments within the block would not result in the intended type of logic.

An **always\_comb** does not need a sensitivity list – the implied sensitivity list includes all signals that are ‘read’ within the block.

An **always\_ff** block requires a sensitivity list that includes **posedge** or **negedge** qualifiers on each signal.

## initial Procedural Blocks

An **initial** block is executed once at the start of the simulation. These are only synthesizable when used to initialize FPGA memory and registers.

### Continuous assignment

An **assign** statement continuously assigns (connects) the result of an expression to a net. It is a more concise way to define combinational logic than using **always\_comb** but is limited to a single expression.

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## Sequential Statements

The following statements appear within **always** or **initial** procedural blocks and execute sequentially (one after the other).

## begin/end

These keywords group statements that should be executed together. They are similar to braces in C. They also begin a new scope for declarations. They can be labelled so that any variables declared within the block can be referenced (e.g. by simulators).

## for/while/do/repeat/forever loops

The **for**, **while** and **do** loops are the same as in C. The **repeat** and **forever** statements execute a statement a given number of times or forever. The **break** and **continue** statements from C can also be used.

loops generate combinational logic and are only synthesizable when the number of iterations is known at compile time. However, they are very useful when writing testbenches for simulation.

## Blocking and Non-Blocking Assignments

A blocking assignment (=) evaluates the RHS (right hand side) and immediately sets the value of the variable on the LHS (left HS).

A non-blocking assignment (<=) evaluates the RHS but does not set the value of the LHS until the next time step (typically, after all sequential statements have executed).

For example,

```
a = b ;  
b = a ;
```

would set both **a** and **b** to the value of **b** while

```
a <= b ;  
b <= a ;
```

would swap the values of **a** and **b**.

Recommended practice is to use non-blocking assignments for sequential logic (in **always\_ff** blocks). This models the behaviour of flip-flops whose outputs don't change until the next rising clock edge.

Blocking assignments are more convenient for designing combinational logic (inside **always\_comb** blocks) as this matches programming language semantics and allows use of intermediate results.

Assignments can synthesize combinational or sequential logic depending on the sensitivity list and

type of the enclosing **always** block as described above.

Do not assign to the same signal from more than one **always** block – the order in which **always** blocks execute is undefined.

You can sometimes break these rules but your code is more likely to have logical errors and the results may not be repeatable (between simulators or from simulation to synthesis).

## if/else

The **if/else** statement syntax is similar to C and synthesizes multiplexers trees whose hierarchy defined by the order of the conditions.

## case/casez

This is the equivalent of C's switch statement. Between **case** and **endcase** are a sequence of expressions, each followed by a colon and a statement.

A **default** value indicates the statement that should be executed if none of the vales match.

By default a case statement synthesizes a multiplexer tree with each condition tested in sequence. Thus the **case** semantics are the same as a sequence of nested **if-else** statements.

Placing **unique** before **case** implies that *exactly one* case expression will match. This allows the expressions to be tested in parallel resulting in faster logic.

Placing **priority** before **case** implies that *at least one* case expression will match and that the first match should be used.

Since both **unique** and **priority** imply that one of the expressions will match, when these are used there should be no **default** expression.

The **casez** variant of the **case** statement allows the case expressions to include ? (or z) values which are treated as “don't care” bits. There is also a **casex** variant where x (undefined) also matches anything which is usually not desired.

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## Other Language Topics

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### Variables and Nets, reg and wire

A *variable* can store a value. It models a register. A variable can only be assigned within an **always** block.

A *net* does not store a value. It models a connection. It must be driven by a continuous assignment statement.

Earlier versions of Verilog used `reg` and `wire` declarations instead of `logic`. Assignments in procedural statements (within `always` blocks) must be to “variables” declared `reg`. Other signals are “nets” and are declared `wire`.

However, `reg` variables need not represent registers and `wire` signals often originate in register outputs. Thus `wire` and `reg` convey little information. Use System Verilog’s `logic` declarations instead whenever possible.

**Exercise 3:** Should each of the following nets (or variables) be declared `wire` or `reg`?

```
module test (a,b,c,d,q) ;
  dff d0 (clk,d,q) ; // assume only q is an output
  assign d = a & b ;
  always@* clk = a & c ;
endmodule
```

Only nets may have multiple drivers (e.g. to model buses with tri-state drivers). There are various flavours (`wire`, `tri`, `wand`, ..) each with different “resolution functions” that combine multiple driving values in different ways. Bidirectional (`inout`) module ports must also be declared with net types.

## Initialization

The ability to initialize registers modeled as ordinary (“static”) variables depends on the target technology. ASIC registers power up in undefined states and must be explicitly reset. However, most CPLD and FPGA technologies allow the power-on values of registers to be included in the configuration data that is loaded when the device is powered on. In this case initialization of static variables is synthesizable.

## System Tasks

Functions beginning with `$` are system tasks. Some examples: `$display()`, similar to C’s `printf()`, can be used to print messages during simulation; `$finish` and `$stop` can terminate a simulation.