

# Capacity and Mutual Information of Wideband Multipath Fading Channels

by I.E. Telatar and D.N.C. Tse (IEEE Trans. IT, July 2000)

proposed by Jan

CTG Reading Group

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## 1 Capacity Bounds for Wideband Multipath Fading Channels

- Basic Assumptions
- Upper Bound on Capacity
- Lower Bound on Capacity

## 2 Mutual Information Achieved by Spread-Spectrum Signaling

- Motivation and Assumptions
- Bounds on Mutual Information
- Practical Implications

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# Capacity Bounds for Wideband Fading Channels I

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- Wideband multipath fading channel, bandwidth  $W$ 
  - ▶ Wideband: received power spread out over large bandwidth
  - ▶ Still narrowband in the sense  $W \ll f_c$  ( $f_c$ : carrier frequency)

- Channel model:

$$y(t) = \sum_{l=1}^L a_l(t) x(t - d_l(t)) + z(t)$$

- ▶  $y(t)$ : received waveform,  $x(t)$ : transmitted waveform,  $z(t)$ : AWGN
- ▶  $L$ : number of physical multipaths
- ▶  $a_l(t)$ : path amplitudes, constant during coherence time  $T_c$ , unknown at receiver
- ▶  $d_l(t)$ : path delays, slowly time-varying, perfectly known at receiver
- Capacity derivation:
  - ▶ Constraint on average received power  $P \Rightarrow \text{SNR} = P/N_0$

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## Upper Bound on Capacity

- Capacity of infinite bandwidth fading channel with SNR  $P/N_0$  and *perfect* channel state information (CSI) at receiver:

$$C^* = \frac{P}{N_0}$$

- Wideband multipath fading channel: path amplitudes  $a_l(t)$  *unknown* at receiver  $\Rightarrow$

$$C \leq C^* = \frac{P}{N_0}$$

- $C^*$  corresponds to capacity of infinite bandwidth AWGN channel (non-fading, perfect CSI at receiver):

$$\lim_{W \rightarrow \infty} W \log \left( 1 + \frac{P}{N_0 W} \right) \approx \lim_{W \rightarrow \infty} W \frac{P}{N_0 W} = \frac{P}{N_0} =: C_{AWGN}$$



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## Lower Bound on Capacity I

Design efficient signaling scheme and assess mutual information (MI)

- Choose symbol duration  $T_s$  such that  $2T_d \leq T_s \leq T_c$   
( $T_d$ : delay spread,  $T_d \ll T_c$  assumed)
- Convey message  $m \in \{1, \dots, M\}$  using signal

$$x_m(t) = \begin{cases} \sqrt{\lambda} \exp(j2\pi f_m t) & 0 \leq t \leq T_s \\ 0 & \text{else} \end{cases}$$

$\Rightarrow$  Single sinusoid at frequency  $f_m$  ( $\hat{=}$  FSK scheme)

- Receiver correlates received signal against all possible  $x_m(t)$ ,  
 $m \in \{1, \dots, M\} \Rightarrow$  *non-coherent* detection
- Choose frequencies as  $f_m := n/(T_s - 2T_d)$  ( $n$  integer) to obtain  
*orthogonal* scheme
- Repeat transmission of  $x_m(t)$  on  $N$  disjoint time intervals  $\Rightarrow$   
receiver can average over fading

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## Lower Bound on Capacity II

- Using *low duty cycle* above scheme achieves MI

$$I(x; y|d_l) = \left(1 - 2\frac{T_d}{T_c}\right) \frac{P}{N_0}$$

Due to *average power constraint* we have  $\lambda := P/\theta \gg P$  ( $\theta \rightarrow 0$ )

- Altogether:

$$\left(1 - 2\frac{T_d}{T_c}\right) C_{AWGN} \leq C \leq C_{AWGN}$$

$$(C_{AWGN} = P/N_0)$$

- ▶ Since  $T_d \ll T_c$ , lower and upper bound approximately *coincide*
- ▶ Capacity-achieving signaling is “peaky” in time and frequency domain

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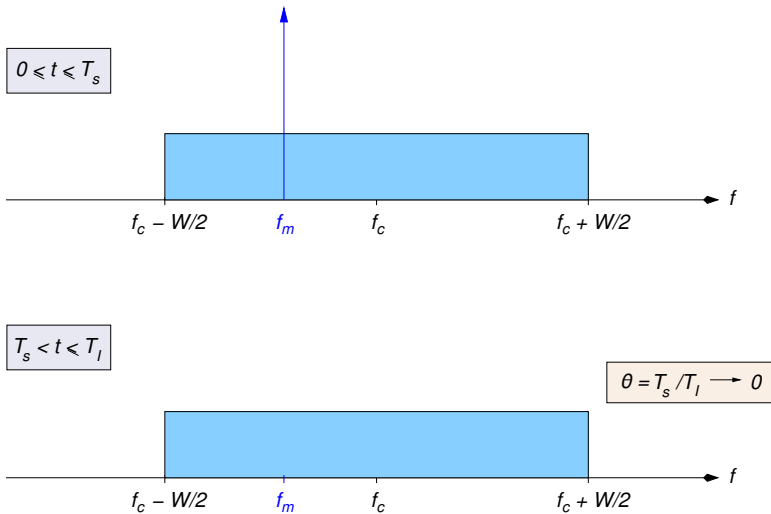
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# MI bounds for Spread-Spectrum Signaling I

*Spread-spectrum (SS)* schemes (DS-CDMA, code-spread CDMA, ...) commonly used for communication over wideband channels

## Key result

- Capacity-achieving signaling for wideband multipath fading channels *maximal different* from SS signaling  
⇒ SS signals are “white-like” and non-peaky in time

## Question

- How good is SS signaling for wideband multipath fading channels?

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## Assumptions

- Discrete-time channel model:

$$Y_i = \sqrt{\frac{\mathcal{E}}{K_c}} \sum_{l=1}^{\tilde{L}} G_l X_{(i-D_l)} + Z_i$$

- ▶  $Y_i$ : received sample,  $X_i$ : transmitted symbol,  $Z_i$ : AWGN sample
  - ▶  $\tilde{L}$ : number of *resolvable* multipaths at system bandwidth  $W$  ( $\tilde{L} \leq L$ )
  - ▶  $G_l, D_l$ : amplitudes/delays of resolvable multipaths
  - ▶  $\mathcal{E} := PT_c/N_0$ ,  $K_c$  normalization factor
- Two different notions of “white-like” signals
    - ▶ info symbols modulated on pseudo-random spreading sequences with near-perfect auto-correlation ( $\hat{=}$  DS-CDMA)
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# MI bounds for Spread-Spectrum Signaling III

## Bounds on MI

- Upper bound on MI per unit time (holds for large  $W$  and large  $\tilde{L}$ ; equal average path energies assumed):

$$I(X; Y|D_I) \leq \frac{\mathcal{E}^2}{T_c^2 \tilde{L}}$$

- Lower bound on MI per unit time (holds for large  $W$  and *any*  $\tilde{L}$ ):

$$I(X; Y|D_I) \geq \frac{\mathcal{E}}{T_c} - \frac{\tilde{L}}{T_c} \log \left( 1 + \frac{\mathcal{E}}{\tilde{L}} \right)$$

- If  $\tilde{L} \ll \mathcal{E}$ , lower bound close to  $\mathcal{E}/T_c = P/N_0 = C_{AWGN}$ , i.e., SS signaling *near-optimal*
- If  $\tilde{L} \gg \mathcal{E}$ , upper bound holds and is close to zero, i.e., SS signaling *highly suboptimal* (!)

$\Rightarrow \mathcal{E} =: \tilde{L}_{crit}$  critical system parameter indicating *overspreading*

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# MI bounds for Spread-Spectrum Signaling IV

## Critical Parameter $\tilde{L}_{crit}$

- Critical parameter also plays *key role* for detection error probability of (specific) binary orthogonal modulation schemes ( $W \rightarrow \infty$ )
- Interpretation of case  $\tilde{L} \gg \tilde{L}_{crit}$ :
  - $\Rightarrow$  energies of resolvable paths very small
  - $\Rightarrow$  poor estimates of complex gains
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